

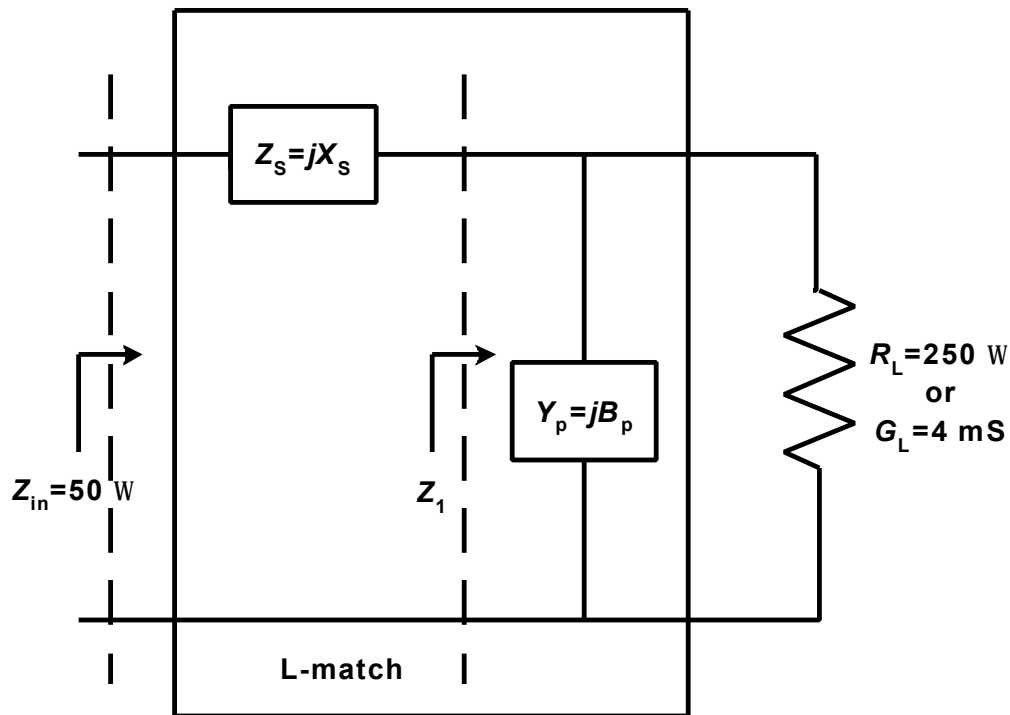
UNIVERSITY ☺ OF ☺ PORTLAND
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L-Matching Network Design

Design Example: Design the L-matching network shown in the figure below to match a purely resistive load with impedance $Z_L = 250 \Omega$ to a purely resistive impedance $Z_{in} = 50 \Omega$ at 1 MHz. (Note that the L-matching network consists of only reactive elements.)



Solution: Combining the admittances $Y_L = Z_L^{-1} = G_L$ and Y_p in parallel ($Y_1 = Y_L + Y_p$) and impedances $Z_1 = Y_1^{-1}$ and Z_S in series ($Z_{in} = Z_1 + Z_S$) yields

$$\underbrace{\left(\frac{1}{Y_L + Y_p} \right)}_{Z_1} + Z_S = \left(\frac{1}{4 \times 10^{-3} + jB_p} \right) + jX_S = Z_{in} = 50 \Omega$$

Using complex conjugate multiplication, this equation can be rewritten as

$$\frac{4 \times 10^{-3} - jB_p}{(4 \times 10^{-3})^2 + B_p^2} + jX_S = \underbrace{\frac{4 \times 10^{-3}}{(4 \times 10^{-3})^2 + B_p^2}}_{\text{Real part}} + j \underbrace{\left(X_S - \frac{B_p}{(4 \times 10^{-3})^2 + B_p^2} \right)}_{\text{Imaginary part}} = 50$$

Equating the real parts of the left- and right-hand sides of this equation, one can solve for B_p , the susceptance value of the parallel element of the L-matching network, as

$$\frac{4 \times 10^{-3}}{(4 \times 10^{-3})^2 + B_p^2} = 50 \rightarrow B_p = \pm 8 \times 10^{-3} \text{ S}$$

The two values found for B_p lead to two different designs. When $B_p = 8 \times 10^{-3} > 0$, the parallel element of the L-matching network is a capacitor with value that can be calculated at the design frequency $f = 1 \text{ MHz}$ as

$$\omega C_p = 2\pi(10^6)C_p = 8 \times 10^{-3} \rightarrow C_p \cong 1.27 \text{ nF} \leftarrow \text{Design \#1}$$

Similarly, when $B_p = -8 \times 10^{-3} < 0$, the parallel element of the L-matching network is an inductor with value that can be obtained as

$$-\frac{1}{\omega L_p} = -\frac{1}{2\pi(10^6)L_p} = -8 \times 10^{-3} \rightarrow L_p \cong 20 \mu\text{H} \leftarrow \text{Design \#2}$$

Equating the imaginary parts and substituting the value of B_p obtained above, one can compute the reactance value of the series element of the L-matching network as

$$X_S - \frac{B_p}{(4 \times 10^{-3})^2 + B_p^2} = 0 \rightarrow X_S = \pm 100 \Omega$$

$B_p = 8 \times 10^{-3} > 0$ leads to $X_S = 100 > 0$, in which case the series element of the L-matching network is an inductor with value given by

$$\omega L_S = 2\pi(10^6)L_S = 100 \rightarrow L_S \cong 15.9 \mu\text{H} \leftarrow \text{Design \#1}$$

Similarly, $B_p = -8 \times 10^{-3} < 0$ leads to $X_S = -100 < 0$, in which case the series element of the L-matching network is a capacitor with a value given by

$$-\frac{1}{\omega C_S} = -\frac{1}{2\pi(10^6)C_S} = -100 \rightarrow C_S \cong 1.59 \text{ nF} \leftarrow \text{Design \#2}$$

The two L-matching networks designed are summarized as shown on the next page. One interesting question that comes to mind is “Is the L-matching network symmetric?” That is instead of converting 250Ω impedance into 50Ω impedance, if we wanted to convert 50Ω impedance into 250Ω impedance, can we use the L-matching networks we already designed to achieve this goal? And if so, how do we connect the circuit? Do we simply change the places of the 250Ω and the 50Ω ? Will this work?

