

For a first-order system with  $V_o(s)/V_i(s) = 1/(\tau s + 1)$ , the relationships are

$$t_r = 2.2\tau = \frac{2.2}{\omega_h} = \frac{0.35}{f_h} \quad (3.51)$$

$$P_o = M_p = 1 \quad (3.52)$$

$$t_p = \infty \quad (3.53)$$

$$t_s = 4\tau \quad (3.54)$$

$$e_1 = \tau \quad (3.55)$$

$$\omega_p = 0 \quad (3.56)$$

For a second-order system with  $V_o(s)/V_i(s) = 1/(s^2/\omega_n^2 + 2\zeta s/\omega_n + 1)$  and  $\theta \triangleq \cos^{-1}\zeta$  (see Fig. 3.7) the relationships are

$$t_r \approx \frac{2.2}{\omega_h} = \frac{0.35}{f_h} \quad (3.57)$$

$$P_o = 1 + \exp \frac{-\pi\zeta}{\sqrt{1-\zeta^2}} = 1 + e^{-\pi/\tan\theta} \quad (3.58)$$

$$t_p = \frac{\pi}{\omega_n \sqrt{1-\zeta^2}} = \frac{\pi}{\omega_n \sin\theta} \quad (3.59)$$

$$t_s \approx \frac{4}{\zeta\omega_n} = \frac{4}{\omega_n \cos\theta} \quad (3.60)$$

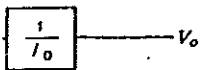
$$e_1 = \frac{2\zeta}{\omega_n} = \frac{2 \cos\theta}{\omega_n} \quad (3.61)$$

$$M_p = \frac{1}{2\zeta \sqrt{1-\zeta^2}} = \frac{1}{\sin 2\theta} \quad \zeta < 0.707, \theta > 45^\circ \quad (3.62)$$

$$\omega_p = \omega_n \sqrt{1-2\zeta^2} = \omega_n \sqrt{-\cos 2\theta} \quad \zeta < 0.707, \theta > 45^\circ \quad (3.63)$$

$$\omega_h = \omega_n (1 - 2\zeta^2 + \sqrt{2 - 4\zeta^2 + 4\zeta^4})^{1/2} \quad (3.64)$$

If a system step response or frequency response is similar to that of an approximating system (see Figs. 3.6, 3.8, 3.11, and 3.12) measurements of  $t_r$ ,  $P_o$ , and  $t_p$  permit estimation of  $\omega_h$ ,  $\omega_p$ , and  $M_p$  or vice versa. The steady-state error in response to a unit ramp can be estimated from either set of measurements.



ps. (a) System with  
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n-step response re-

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nse  $M_p$ .

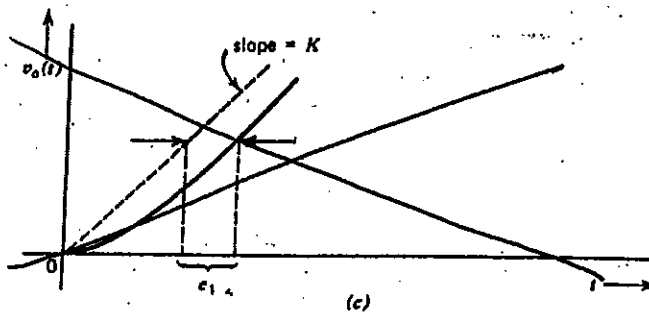
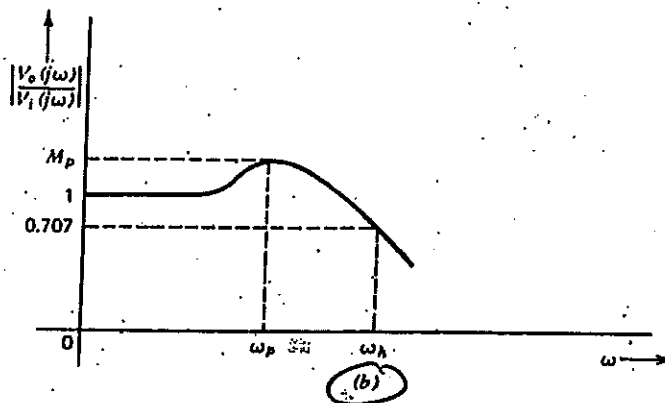
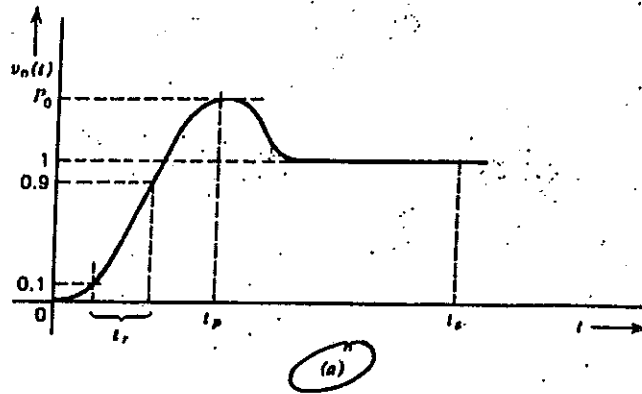


Figure 3.17 Parameters used to describe transient and frequency responses. (a) Unit-step response. (b) Frequency response. (c) Ramp response.

One final c  
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3.6 ERRO

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Radar  
antenna

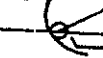


Figure 3.18 Radar antenna